

## CONNECTING ELECTRICAL CIRCUITS: TRANSVERSALITY AND WELL-POSEDNESS

By

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### 1. Introduction.

An electrical circuit  $\mathcal{E}$  consists of three kinds of elements, inductors, capacitors and resistors mutually connected. A state of an electrical circuit is specified by a current vector  $\mathbf{i}=(i_1, i_2, \dots, i_b) \in R^b$  and a voltages vector  $\mathbf{v}=(v_1, v_2, \dots, v_b) \in R^b$ . Let  $G$  be the (oriented) graph determined naturally by the circuit  $\mathcal{E}$ . We can regard  $\mathbf{i}$  and  $\mathbf{v}$  as a real 1-chain and 1-cochain of  $G$ , i.e.,  $\mathbf{i} \in C_1(G)$  and  $\mathbf{v} \in C^1(G)$ .\* The Kirchhoff current (voltage) law restricts  $\mathbf{i}(\mathbf{v})$  to belong to a linear subspace  $K_c(\mathcal{E})=\text{Ker } \partial \subset C_1(G)$  ( $K_v(\mathcal{E})=\text{Im } \partial^* \subset C^1(G)$ ), where  $\partial: C_1(G) \rightarrow C_0(G)$  ( $\partial^*: C^0(G) \rightarrow C^1(G)$ ) is boundary (coboundary) operator. Another restriction of possible states is the restraint of resistive characteristics. We admit couplings between same kind of elements. The resistive characteristics are represented by a  $\rho$ -dimensional smooth submanifold  $A_R \subset C_1(G_R) \times C^1(G_R)$ , where  $\rho$  is the number of resistive elements and  $C_1(G_R)(C^1(G_R))$  is  $\rho$ -dimensional euclidian space consisting of resistive currents (voltages).

Thus the currents and voltages  $(\mathbf{i}, \mathbf{v})=(\mathbf{i}_L, \mathbf{i}_C, \mathbf{i}_R, \mathbf{v}_L, \mathbf{v}_C, \mathbf{v}_R)$  must belong to the configuration space;

$$\Sigma = K \cap A, \quad K = K_c(G) \times K_v(G), \quad A = \pi_R^{-1}(A_R),$$

where  $\pi_R: C_1(G) \times C^1(G) \rightarrow C_1(G_R) \times C^1(G_R)$  is the natural projection, i.e.,  $\pi_R(\mathbf{i}_L, \mathbf{i}_C, \mathbf{i}_R, \mathbf{v}_L, \mathbf{v}_C, \mathbf{v}_R) = (\mathbf{i}_R, \mathbf{v}_R)$ .

For dynamics to be defined on whole of  $\Sigma$ , two things are needed. One is the transversality of  $K$  and  $A$  which assures  $\Sigma$  to be smooth submanifold. Since  $K$  is  $b$ -dimensional,  $\dim \Sigma = 2b - (b + \rho) = b - \rho$  which is equal to the number of inductors and capacitors. Another is the regularity of the map

$$\pi_{LC}: \Sigma \rightarrow C_1(G_L) \times C^1(G_C),$$

which is the restriction of the natural projection to  $\Sigma$ , i.e.,

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\* It is no suffering, however, to regard simply  $C_1(G)$  and  $C^1(G)$  as  $b$ -dimensional euclidian space in the sequel.

$$\pi_{LC}(\mathbf{i}_L, \mathbf{i}_C, \mathbf{i}_R, \mathbf{v}_L, \mathbf{v}_C, \mathbf{v}_R) = (\mathbf{i}_L, \mathbf{v}_C).$$

If  $D\pi_{LC}(x)$  is non-singular at any point  $x \in \Sigma$ , we can determine the smooth vector field  $X$  on  $\Sigma$  which describes the dynamics of the circuit ([5], [3], [2]). Then any solution curve is locally a solution of certain differential equation. Therefore so called 'jumping phenomena' does not occur. We call such a circuit *well-posed*.

Consider a circuit  $\mathcal{C}$  whose graph  $G = G(\mathcal{C})$  has a proper tree. Recall a *proper tree* means a tree which contains all the capacitors and no inductor. Since in a circuit without proper tree the map  $\pi_{LC}$  is singular at any point of  $\Sigma$ , we assume the existence of proper tree ([3]). To assure  $\Sigma$  to be a submanifold, we need the transversality of the characteristic submanifold  $\Lambda$  and the Kirchhoff space  $K$ . Now, we define a somewhat stronger condition than the transversality condition. We say  $\Lambda$  and  $K$  are *everywhere transverse* if for every  $x \in \Lambda$ ,  $T_x(\Lambda)$  is transverse to  $K$  in  $C_1(G) \times C^1(G)$ . We call such a circuit itself everywhere transverse circuit. In the same way, we will define a notion of *strongly well-posedness* in section 3, these notions play a central role in this paper.

Now, we consider a connection of two circuits by two different kinds of (uncoupled) elements. For given circuits  $\mathcal{C}_1$  and  $\mathcal{C}_2$ , connecting a node  $p$  of  $\mathcal{C}_1$  with a node  $r$  of  $\mathcal{C}_2$  by an uncoupled element  $e_1$  and connecting a node  $q$  of  $\mathcal{C}_1$  with a node  $s$  of  $\mathcal{C}_2$  by an uncoupled element  $e_2$ , we obtain a new circuit  $\mathcal{C}$ .

We will discuss in §2 the transversality problem, in §3 the well-posedness problem of the new circuit  $\mathcal{C}$  obtained by connecting two circuits as stated above and in §4 some examples of strongly well-posed circuits. For the reason of connecting by different kinds of elements, see §4.

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## 2. Transversality.

Let  $\mathcal{C}$  be the circuit obtained by connecting two circuits  $\mathcal{C}_i$ ,  $i=1, 2$ , by two uncoupled elements  $e_1$  and  $e_2$  as stated before.

**Theorem 1.** *Suppose the circuit  $\mathcal{C}_i$  is everywhere transverse,  $i=1, 2$ . Then the new circuit  $\mathcal{C}$  is also everywhere transverse.*

We say a node  $p$  is *L-connected* with  $q$  in  $\mathcal{C}_1$  if there exists a path from  $p$  to  $q$  in  $\mathcal{C}_1$  which is consisting of only inductor branches.

**Theorem 2.** *Suppose the circuit  $\mathcal{C}_i$  is transverse,  $i=1, 2$ ,  $p$  is L-connected with  $q$  in  $\mathcal{C}_1$  and  $r$  is L-connected with  $s$  in  $\mathcal{C}_2$ . Then the new circuit  $\mathcal{C}$  is also transverse.*

**Proof of Theorem 1.** First, we deal with the following case.

**Case 1.** The element  $e_1$  is a capacitor and  $e_2$  is an inductor.

Let  $\mathcal{T}_i$  be a tree for  $\mathcal{E}_i$ ,  $i=1, 2$ . Put  $\mathcal{T} = \mathcal{T}_1 \cup \{e_1\} \cup \mathcal{T}_2$  and clearly  $\mathcal{T}$  is a tree for  $G(\mathcal{E})$ . Let  $C_1(\mathcal{E}_j)$  and  $C^1(\mathcal{E}_j)$  denote the current and voltage spaces of  $\mathcal{E}_j$ , respectively,  $j=1, 2$ . Since we do not add a new resistive element, the characteristic submanifold of the new circuit  $\mathcal{E}$  is the following;

$$(1) \quad \Lambda(\mathcal{E}) = \Lambda(\mathcal{E}_1) \times \Lambda(\mathcal{E}_2) \times C_1(e_1) \times C^1(e_1) \times C_1(e_2) \times C^1(e_2) \subset C_1(G(\mathcal{E})) \times C^1(G(\mathcal{E})).$$

The Kirchhoff space  $K(\mathcal{E})$  of  $\mathcal{E}$  is represented as follows. Let  $B_j$  denote the fundamental loop matrix for  $\mathcal{E}_j$  corresponding to the tree  $\mathcal{T}_j$ ,  $j=1, 2$ . Then the fundamental loop matrix  $B$  for corresponding to  $\mathcal{T}$  has the following form;

$$(2) \quad B = [I: A] = \begin{bmatrix} \mathcal{L}_1 & \mathcal{L}_2 & e_2 & \mathcal{T}_1 & \mathcal{T}_2 & e_1 \\ I & & & A_1 & & \\ & I & & & A_2 & \\ & & 1 & A_{21} & A_{22} & -1 \end{bmatrix} \begin{matrix} \mathcal{L}_1 \\ \mathcal{L}_2 \\ e_2 \end{matrix},$$

where  $B_j = [I: A_j]$  and  $A_{21}(A_{12})$  has non-zero element ( $\pm 1$ ) at the tree branches in  $\mathcal{E}_1(\mathcal{E}_2)$  contained in the path connecting  $p(r)$  with  $q(s)$  in  $\mathcal{T}_1(\mathcal{T}_2)$ . The current Kirchhoff space  $K_c(\mathcal{E})$  of  $\mathcal{E}$  is the image of the map

$$B^t: C_1(\mathcal{L}) \rightarrow C_1(G(\mathcal{E})),$$

where  $C_1(\mathcal{L})$  denotes the current space of the link branches in  $\mathcal{E}$ . And the voltage Kirchhoff space  $K_v(\mathcal{E})$  of  $\mathcal{E}$  is the image of the map

$$Q^t: C^1(\mathcal{T}) \rightarrow C^1(G(\mathcal{E})),$$

where  $Q$  is the fundamental cutset matrix for  $\mathcal{E}$  and has the form  $([4], [1])$ ;

$$(3) \quad Q = [-A^t: I],$$

and  $C^1(\mathcal{T})$  denotes the voltage space of the tree branches in  $\mathcal{E}$ . Thus the Kirchhoff space  $K(\mathcal{E})$  of  $\mathcal{E}$  is represented as follows;

$$(4) \quad K(\mathcal{E}) = \text{Im } B^t \times \text{Im } Q^t \subset C_1(G(\mathcal{E})) \times C^1(G(\mathcal{E})).$$

Now, we prove the everywhere transversality of  $\Lambda(\mathcal{E})$  and  $K(\mathcal{E})$ . Take a point  $(i, v) \in \Lambda(\mathcal{E})$ . We must show that  $T_{(i, v)}\Lambda(\mathcal{E})$  is transverse to  $K(\mathcal{E})$ . To show this, for each vector  $w \in T_{(i, v)}C_1(G(\mathcal{E})) \times C^1(G(\mathcal{E}))$  we will find two vectors  $w_A \in T_{(i, v)}\Lambda(\mathcal{E})$  and  $w_K \in K(\mathcal{E})$  such that  $w = w_A + w_K$ .

Corresponding to the decomposition;

$$\begin{aligned}
C_1(G(\mathcal{E})) \times C^1(G(\mathcal{E})) &= C_1(G(\mathcal{E}_1)) \times C^1(G(\mathcal{E}_1)) \times C_1(G(\mathcal{E}_2)) \times C^1(G(\mathcal{E}_2)) \times C_1(e_1) \\
&\quad \times C^1(e_1) \times C_1(e_2) \times C^1(e_2), \\
&= C_1(\mathcal{T}_1) \times C_1(\mathcal{T}_2) \times C_1(\mathcal{L}_1) \times C_1(\mathcal{L}_2) \times C_1(e_1) \times C_1(e_2) \\
&\quad \times C^1(\mathcal{T}_1) \times C^1(\mathcal{T}_2) \times C^1(\mathcal{L}_1) \times C^1(\mathcal{L}_2) \times C^1(e_1) \times C^1(e_2),
\end{aligned}$$

we write

$$\begin{aligned}
(\mathbf{i}, \mathbf{v}) &= (\mathbf{i}_1, \mathbf{v}_1, \mathbf{i}_2, \mathbf{v}_2, \mathbf{i}_{e_1}, \mathbf{v}_{e_1}, \mathbf{i}_{e_2}, \mathbf{v}_{e_2}) \\
&= (\mathbf{i}_{\mathcal{T}_1}, \mathbf{i}_{\mathcal{T}_2}, \mathbf{i}_{\mathcal{L}_1}, \mathbf{i}_{\mathcal{L}_2}, \mathbf{i}_{e_1}, \mathbf{i}_{e_2}, \mathbf{v}_{\mathcal{T}_1}, \mathbf{v}_{\mathcal{T}_2}, \mathbf{v}_{\mathcal{L}_1}, \mathbf{v}_{\mathcal{L}_2}, \mathbf{v}_{e_1}, \mathbf{v}_{e_2})
\end{aligned}$$

and

$$\begin{aligned}
\mathbf{w} &= (\mathbf{w}_1, \mathbf{w}_2, \mathbf{w}_{e_1}, \mathbf{w}_{e_2}) \\
&= (\mathbf{w}(\mathbf{i}_{\mathcal{T}_1}), \mathbf{w}(\mathbf{i}_{\mathcal{T}_2}), \mathbf{w}(\mathbf{i}_{\mathcal{L}_1}), \mathbf{w}(\mathbf{i}_{\mathcal{L}_2}), \mathbf{w}(\mathbf{i}_{e_1}), \mathbf{w}(\mathbf{i}_{e_2}), \\
&\quad \mathbf{w}(\mathbf{v}_{\mathcal{T}_1}), \mathbf{w}(\mathbf{v}_{\mathcal{T}_2}), \mathbf{w}(\mathbf{v}_{\mathcal{L}_1}), \mathbf{w}(\mathbf{v}_{\mathcal{L}_2}), \mathbf{w}(\mathbf{v}_{e_1}), \mathbf{w}(\mathbf{v}_{e_2})).
\end{aligned}$$

Then by (1),  $(\mathbf{i}_j, \mathbf{v}_j) \in \Lambda(\mathcal{E})$ ,  $j=1, 2$ . Since  $\Lambda(\mathcal{E}_j)$  is everywhere transverse to  $K(\mathcal{E}_j)$ , there exist vectors  $\mathbf{w}_{\Lambda_j} \in T_{(\mathbf{i}_j, \mathbf{v}_j)}\Lambda(\mathcal{E})$  and  $\mathbf{w}_{K_j} \in K(\mathcal{E}_j) = T_{(\mathbf{i}_j, \mathbf{v}_j)}(K(\mathcal{E}_j) + (\mathbf{i}_j, \mathbf{v}_j))$  such that  $\mathbf{w}_j = \mathbf{w}_{K_j} + \mathbf{w}_{\Lambda_j}$ ,  $j=1, 2$ . Put  $\mathbf{w}_{\Lambda} = (\mathbf{w}_{\Lambda_1}, \mathbf{w}_{\Lambda_2}, \mathbf{w}_{\Lambda}(\mathbf{i}_{e_1}), \mathbf{w}_{\Lambda}(\mathbf{i}_{e_2}), \mathbf{w}_{\Lambda}(\mathbf{v}_{e_1}), \mathbf{w}_{\Lambda}(\mathbf{v}_{e_2}))$  then  $\mathbf{w}_{\Lambda} \in T_{(\mathbf{i}, \mathbf{v})}\Lambda(\mathcal{E})$  for any  $\mathbf{w}(\mathbf{i}_{e_j})$  and  $\mathbf{w}(\mathbf{v}_{e_j})$ ,  $j=1, 2$ , by (1). Put  $\mathbf{w}_K = (\mathbf{w}_{K_1}, \mathbf{w}_{K_2}, \mathbf{w}_K(\mathbf{i}_{e_1}), \mathbf{w}_K(\mathbf{i}_{e_2}), \mathbf{w}_K(\mathbf{v}_{e_1}), \mathbf{w}_K(\mathbf{v}_{e_2}))$  and decompose,

$$\mathbf{w}_{K_j} = (\mathbf{w}_{K_j}(\mathbf{i}), \mathbf{w}_{K_j}(\mathbf{v})), \quad j=1, 2,$$

and

$$\begin{aligned}
\mathbf{w}_{K_j}(\mathbf{i}) &= (\mathbf{w}_K(\mathbf{i}_{\mathcal{T}_j}), \mathbf{w}_K(\mathbf{i}_{\mathcal{L}_j})), \quad j=1, 2, \\
\mathbf{w}_{K_j}(\mathbf{v}) &= (\mathbf{w}_K(\mathbf{v}_{\mathcal{T}_j}), \mathbf{w}_K(\mathbf{v}_{\mathcal{L}_j})), \quad j=1, 2.
\end{aligned}$$

Then by (2), (3) and (4),  $\mathbf{w}_K$  belongs to  $K(\mathcal{E})$  if and only if

$$\begin{aligned}
\mathbf{w}_K(\mathbf{i}_{\mathcal{T}_j}) &= A_j^t \mathbf{w}_K(\mathbf{i}_{\mathcal{L}_j}) + A_{2j}^t \mathbf{w}_K(\mathbf{i}_{e_2}), \quad j=1, 2, \\
\mathbf{w}_K(\mathbf{i}_{e_1}) &= -\mathbf{w}_K(\mathbf{i}_{e_2}),
\end{aligned}$$

(5) and

$$\begin{aligned}
\mathbf{w}_K(\mathbf{v}_{\mathcal{L}_j}) &= -A_j \mathbf{w}_K(\mathbf{v}_{\mathcal{T}_j}), \quad j=1, 2, \\
\mathbf{w}_K(\mathbf{v}_{e_2}) &= -A_{21} \mathbf{w}_K(\mathbf{v}_{\mathcal{T}_1}) - A_{22} \mathbf{w}_K(\mathbf{v}_{\mathcal{T}_2}) + \mathbf{w}_K(\mathbf{v}_{e_1}).
\end{aligned}$$

Thus, if we put  $\mathbf{w}_K(\mathbf{i}_{e_2}) = \mathbf{w}_K(\mathbf{i}_{e_1}) = 0$ ,  $\mathbf{w}_K(\mathbf{v}_{e_1}) = 0$ ,  $\mathbf{w}_K(\mathbf{v}_{e_2}) = -A_{21} \mathbf{w}_K(\mathbf{v}_{\mathcal{T}_1}) - A_{22} \mathbf{w}_K(\mathbf{v}_{\mathcal{T}_2})$ ,  $\mathbf{w}_{\Lambda}(\mathbf{i}_{e_1}) = \mathbf{w}(\mathbf{i}_{e_1})$ ,  $\mathbf{w}_{\Lambda}(\mathbf{i}_{e_2}) = \mathbf{w}(\mathbf{i}_{e_2})$ ,  $\mathbf{w}_{\Lambda}(\mathbf{v}_{e_1}) = \mathbf{w}(\mathbf{v}_{e_1})$ , and  $\mathbf{w}_{\Lambda}(\mathbf{v}_{e_2}) = \mathbf{w}(\mathbf{v}_{e_2}) - \mathbf{w}_K(\mathbf{v}_{e_2})$ , then  $\mathbf{w} = \mathbf{w}_{\Lambda} + \mathbf{w}_K$  and  $\mathbf{w}_{\Lambda} \in T_{(\mathbf{i}, \mathbf{v})}\Lambda(\mathcal{E})$ ,  $\mathbf{w}_K \in K(\mathcal{E})$ . This proves the everywhere transversality of  $\Lambda(\mathcal{E})$  and  $K(\mathcal{E})$ .

**Case 2.** The element  $e_1$  is a linear resistor and the element  $e_2$  is an inductor or a capacitor.

Put  $\mathcal{T} = \mathcal{T}_1 \cup \{e_1\} \cup \mathcal{T}_2$  and  $\mathcal{T}$  is a tree. Let  $\Lambda(e_1)$  denote the characteristics of  $e_1$ . Since  $e_1$  is a linear resistor, we have

$$\Lambda(e_1) = \{(i_{e_1}, v_{e_1}) : v_{e_1} = i_{e_1} \cdot r_{e_1}\} \subset C_1(e_1) \times C^1(e_1),$$

The characteristic submanifold  $\Lambda(\mathcal{E})$  of the new circuit has the following form;

$$\Lambda(\mathcal{E}) = \Lambda(\mathcal{E}_1) \times \Lambda(\mathcal{E}_2) \times \Lambda(e_1).$$

The Kirchoff space  $K(\mathcal{E})$  is quite the same as in case 1. For a point  $(i, v) \in \Lambda(\mathcal{E})$ , we will show the transversality of  $T_{(i,v)}\Lambda(\mathcal{E})$  and  $K(\mathcal{E})$ . As in case 1, for each  $w \in T_{(i,v)}(C_1(G(\mathcal{E})) \times C^1(G(\mathcal{E})))$  we must find  $w_A \in T_{(i,v)}\Lambda(\mathcal{E})$  and  $w_K \in K(\mathcal{E})$  such that  $w = w_K + w_A$ . The only difference from case 1 is that the vector  $(w_A(i_{e_1}), w_A(v_{e_1}))$  should belong to  $T_{(i_{e_1}, v_{e_1})}\Lambda(e_1) = \Lambda(e_1)$ , i.e.,  $w_A(i_{e_1}) \cdot r_{e_1} = w_A(v_{e_1})$ . Besides (5) put

$$w_A(i_{e_1}) = w(i_{e_1}), \quad w_A(v_{e_1}) = w_A(i_{e_1}) \cdot r_{e_1}, \quad w_K(v_{e_1}) = w(v_{e_1}) - w_A(v_{e_1}),$$

and

$$w_A(v_{e_2}) = w(v_{e_2}) - w_K(v_{e_2}).$$

Then  $w_K$  and  $w_A$  have the required properties.

**Proof of Theorem 2.** The proof of Theorem 2 proceeds in the same way as that of Theorem 1 except one point that if  $(i_j, v_j)$  does not belong to  $K(\mathcal{E}_j)$  then  $T_{(i_j, v_j)}\Lambda(\mathcal{E}_j)$  may not transverse to  $K(\mathcal{E}_j)$ . This point is rescued by the assumption of  $L$ -connectedness as follows.

Since  $(i, v)$  belongs to  $K(\mathcal{E})$ ,  $(i, v)$  satisfies (5) or

$$\begin{aligned} i_{\mathcal{L}_j} &= A_j^t i_{\mathcal{L}_j} + A_{2j}^t i_{e_2}, \quad j=1, 2, \\ i_{e_1} &= -i_{e_2}, \\ v_{\mathcal{L}_j} &= -A_j v_{\mathcal{L}_j}, \quad j=1, 2, \\ v_{e_2} &= -A_{21} v_{\mathcal{L}_1} - A_{22} v_{\mathcal{L}_2} + v_{e_1}. \end{aligned}$$

The assumption that  $p$  and  $q$  are connected by the path consisting of only inductors means the following. There exists a link currents vector  $\hat{i}_{\mathcal{L}_j} \in C_1(G(\mathcal{L}_j))$  such that  $i_{\mathcal{L}_j} = A_j \hat{i}_{\mathcal{L}_j}$ ,  $j=1, 2$ , and the difference between  $i_{\mathcal{L}_1}(i_{e_2})$  and  $\hat{i}_{\mathcal{L}_1}(\hat{i}_{e_2})$  occurs only in the link inductor branches connecting  $p(r)$  with  $q(s)$  in  $\mathcal{E}_1(\mathcal{E}_2)$ . If we put  $\hat{i}_{\mathcal{L}_j} = (\hat{i}_{\mathcal{L}(L)_j}, \hat{i}_{\mathcal{L}(R)_j})$  then  $\hat{i}_{\mathcal{L}_j} = (\hat{i}_{\mathcal{L}(L)_j}, \hat{i}_{\mathcal{L}(R)_j})$ . Put  $(i'_j, v_j) = (i_{\mathcal{L}_j}, \hat{i}_{\mathcal{L}_j}, v_{\mathcal{L}_j}, v_{\mathcal{L}_j})$  and  $(i'_j, v_j) \in K(\mathcal{E}_j)$  and  $(i'_j, v_j) \in \Lambda(\mathcal{E}_j)$ . Since there is no coupling between different kinds of elements,  $\Lambda(\mathcal{E}_j)$  has the form;  $\Lambda(\mathcal{E}_j) = \pi_{R_j}^{-1}(A_{R_j})$ ,  $j=1, 2$ , where  $\pi_{R_j}: C_1(G(\mathcal{E}_j)) \times C^1(G(\mathcal{E}_j)) \rightarrow C_1(G_R(\mathcal{E}_j)) \times C^1(G_R(\mathcal{E}_j))$  is the natural projection to the resistive current and voltage space,  $j=1, 2$ . Therefore  $T_{(i'_j, v_j)}\Lambda(\mathcal{E}_j) = T_{(i_j, v_j)}\Lambda(\mathcal{E}_j)$ . By the assumption  $T_{(i'_j, v_j)}\Lambda(\mathcal{E}_j)$  is transverse to  $K(\mathcal{E}_j)$ . Thus  $T_{(i_j, v_j)}\Lambda(\mathcal{E}_j)$  is transverse to  $K(\mathcal{E}_j)$ . The rest of the proof is the same as that of Theorem 1.

### 3. Well-posedness.

We will discuss the problem when the connected circuit is well-posed. Henceforth, we assume all the circuits have proper trees and are transverse. First we show a variant of Theorem in [2], which gives a necessary and sufficient condition for a circuit to be well-posed in terms of transversality. Let us recall some notations from [2]. For  $(\mathbf{i}_L, \mathbf{v}_C) \in C_1(G_L) \times C^1(G_C)$ , put

$$K(\mathbf{i}_L, \mathbf{v}_C) = \pi_{LC}^{-1}(\mathbf{i}_L, \mathbf{v}_C) \cap K,$$

where  $\pi_{LC}: C_1(G) \times C^1(G) \rightarrow C_1(G_L) \times C^1(G_C)$  is the natural projection and  $K$  is the Kirchhoff space. By (4), we can see the space  $K(\mathbf{i}_L, \mathbf{v}_C)$  is parallel translation of the space  $K(0, 0)$  to a point  $b(\mathbf{i}_L, \mathbf{v}_C)$  in  $C_1(G) \times C^1(G)$ . Here,  $b(\mathbf{i}_L, \mathbf{v}_C) = \begin{bmatrix} B^t & 0 \\ 0 & Q^t \end{bmatrix} \begin{bmatrix} 0 \\ \mathbf{i}_L \\ 0 \\ \mathbf{v}_C \end{bmatrix}$ .

Since  $\pi_R: C_1(G) \times C^1(G) \rightarrow C_1(G_R) \times C^1(G_R)$  is a linear map, the space  $\pi_R(K(\mathbf{i}_L, \mathbf{v}_C))$  is also the parallel translation of  $K_0 = \pi_R(K(0, 0))$  to the point  $(\mathbf{i}_R, \mathbf{v}_R) = \pi_R(b(\mathbf{i}_L, \mathbf{v}_C))$ .

**Theorem A.** *A circuit  $\mathcal{C}$  is well-posed if and only if the affine subspace  $\pi_R(K(\mathbf{i}_L, \mathbf{v}_C))$  is transverse to the characteristic submanifold  $A_R$  for all  $(\mathbf{i}_L, \mathbf{v}_C) \in C_1(G_L) \times C^1(G_C)$ .*

**Proof of Theorem A.** Assume  $\mathcal{C}$  is not well-posed. Then there exist a singular point  $(\mathbf{i}, \mathbf{v}) = (\mathbf{i}_L, \mathbf{i}_C, \mathbf{i}_R, \mathbf{v}_L, \mathbf{v}_C, \mathbf{v}_R) \in \Sigma$ . By Theorem in [2], the space  $K_0 + (\mathbf{i}_R, \mathbf{v}_R) = \pi_R(K(\mathbf{i}_L, \mathbf{v}_C))$  is not transverse to  $A_R$  at  $(\mathbf{i}_R, \mathbf{v}_R)$ . This proves the 'if' part of Theorem A.

Let  $\mathcal{T}$  be a proper tree and  $\mathcal{L}$  a corresponding link. Assume  $(\mathbf{i}_L, \mathbf{v}_C) \in C_1(G_L) \times C^1(G_C)$  be a point such that  $\pi_R(K(\mathbf{i}_L, \mathbf{v}_C))$  and  $A_R$  have non-transversal intersection. Let  $(\mathbf{i}_R, \mathbf{v}_R) = (\mathbf{i}_{R(\mathcal{T})}, \mathbf{i}_{R(\mathcal{L})}, \mathbf{v}_{R(\mathcal{T})}, \mathbf{v}_{R(\mathcal{L})}) \in \pi_R(K(\mathbf{i}_L, \mathbf{v}_C)) \cap A_R$  be a non-transversal point, where  $\mathbf{i}_{R(\mathcal{T})}$  and  $\mathbf{i}_{R(\mathcal{L})}$  ( $\mathbf{v}_{R(\mathcal{T})}$  and  $\mathbf{v}_{R(\mathcal{L})}$ ) denote the currents (voltages) of tree resistors and link resistors, respectively. Since  $(\mathbf{i}_R, \mathbf{v}_R)$  belongs to  $\pi_R(K(\mathbf{i}_L, \mathbf{v}_C))$ , the following holds

$$\begin{aligned} \mathbf{i}_{R(\mathcal{T})} &= A_{R\mathcal{T}}^t \mathbf{i}_{R(\mathcal{L})} + A_{L\mathcal{T}}^t \mathbf{i}_L, \\ \mathbf{v}_{R(\mathcal{L})} &= -A_{R\mathcal{L}} \mathbf{v}_{R(\mathcal{T})} - A_{RC} \mathbf{v}_C, \end{aligned}$$

where  $A_{R\mathcal{T}}$ ,  $A_{L\mathcal{T}}$  and  $A_{RC}$  are the submatrices of  $B$  given by the following form;

$$(6) \quad B = \begin{bmatrix} R(\mathcal{L}) & L & R(\mathcal{T}) & C \\ 1 & 0 & A_{R\mathcal{T}} & A_{RC} \\ 0 & 1 & A_{L\mathcal{T}} & A_{LC} \end{bmatrix} \begin{bmatrix} R(\mathcal{L}) \\ L \end{bmatrix}.$$

Put

$$\begin{aligned} \mathbf{i}_C &= A_{RC}^t \mathbf{i}_{R(\mathcal{I})} + A_{LC}^t \mathbf{i}_L, \\ \mathbf{v}_L &= -A_{L\mathcal{I}} \mathbf{v}_{R(\mathcal{I})} - A_{LC} \mathbf{v}_C. \end{aligned}$$

Then

$$\begin{aligned} \begin{bmatrix} \mathbf{i}_{R(\mathcal{I})} \\ \mathbf{i}_C \end{bmatrix} &= A^t \begin{bmatrix} \mathbf{i}_{R(\mathcal{I})} \\ \mathbf{i}_L \end{bmatrix}, \\ \begin{bmatrix} \mathbf{v}_{R(\mathcal{I})} \\ \mathbf{v}_L \end{bmatrix} &= -A \begin{bmatrix} \mathbf{v}_{R(\mathcal{I})} \\ \mathbf{v}_C \end{bmatrix}, \end{aligned}$$

this means

$$(\mathbf{i}, \mathbf{v}) = (\mathbf{i}_L, \mathbf{i}_C, \mathbf{i}_R, \mathbf{v}_L, \mathbf{v}_C, \mathbf{v}_R) \in K(\mathbf{i}_L, \mathbf{v}_C).$$

Since  $(\mathbf{i}_R, \mathbf{v}_R)$  belongs to  $\Lambda_R$ ,  $(\mathbf{i}, \mathbf{v})$  belongs to  $\Lambda = \pi_R^{-1}(\Lambda_R)$ . Applying Theorem in [2] for a point  $(\mathbf{i}, \mathbf{v}) \in K(\mathbf{i}_L, \mathbf{v}_C) \cap \Lambda \subset K \cap \Lambda = \Sigma$ , we see that  $(\mathbf{i}, \mathbf{v})$  is a singular point. This proves the 'only if' part of Theorem A.

According to Theorem A, we can reduce the well-posedness problem to the transversality problem. Now we define a stronger condition than well-posedness, corresponding to the 'everywhere transversality' in section 2. A circuit is called *strongly well-posed* if for all  $(\mathbf{i}_R, \mathbf{v}_R) \in \Lambda_R$ ,  $T_{(\mathbf{i}_R, \mathbf{v}_R)} \Lambda_R$  is transverse to  $K_0$ .

**Theorem 3.** *Suppose the circuit  $\mathcal{E}_i$  is strongly well-posed,  $i=1, 2$ . Then the new circuit  $\mathcal{E}$  is also strongly well-posed.*

**Theorem 4.** *Suppose the circuit  $\mathcal{E}_i$  is well-posed,  $i=1, 2$ . If  $p(r)$  is  $L$ -connected with  $q(s)$  in  $\mathcal{E}_1(\mathcal{E}_2)$ , then the new circuit  $\mathcal{E}$  is also well-posed.*

**Proof of Theorem 3.** This is essentially the same as that of Theorem 1.

**Case 1.** The element  $e_1$  is a capacitor and  $e_2$  is an inductor.

We can see the following holds by direct verification or by noting that the space  $K_0$  agrees with the Kirchhoff space of the resistive circuit obtained by open-circuiting the inductor branches and short-circuiting the capacitor branches,

$$(7) \quad K_0(\mathcal{E}) = K_0(\mathcal{E}_1) \times K_0(\mathcal{E}_2).$$

Clearly,

$$(8) \quad \Lambda_R(\mathcal{E}) = \Lambda_R(\mathcal{E}_1) \times \Lambda_R(\mathcal{E}_2) \subset C_1(G_R(\mathcal{E})) \times C^1(G_R(\mathcal{E})).$$

Take a point  $(\mathbf{i}_R, \mathbf{v}_R) \in \Lambda_R$ . Put

$$(\mathbf{i}_R, \mathbf{v}_R) = (\mathbf{i}_{R_1}, \mathbf{v}_{R_1}, \mathbf{i}_{R_2}, \mathbf{v}_{R_2}) \in C_1(G_R(\mathcal{E}_1)) \times C^1(G_R(\mathcal{E}_1)) \times C_1(G_R(\mathcal{E}_2)) \times C^1(G_R(\mathcal{E}_2))$$

and by (8)  $(\mathbf{i}_{R_j}, \mathbf{v}_{R_j}) \in \Lambda_R(\mathcal{E}_j)$ . Since  $\mathcal{E}_j$  is strongly well-posed,  $T_{(\mathbf{i}_{R_j}, \mathbf{v}_{R_j})}(\Lambda_R(\mathcal{E}_j))$  and  $K_0(\mathcal{E}_j)$  are transverse by (7) and (8).

**Case 2.** The element  $e_1$  is a linear resistor and the element  $e_2$  is an inductor. In this case, we have

$$(9) \quad A_R(\mathcal{E}) = A_R(\mathcal{E}_1) \times A_R(\mathcal{E}_2) \times A_R(e_1)$$

where

$$A_R(e_1) = \{(i_{e_1}, v_{e_1}) : v_{e_1} = i_{e_1} \cdot r_{e_1}\} \subset C_1(e_1) \times C^1(e_1).$$

And

$$(10) \quad K_0(\mathcal{E}) = K_0(\mathcal{E}_1) \times K_0(\mathcal{E}_2) \times K_0(e_1),$$

where

$$K_0(e_1) = \{(i_{e_1}, v_{e_1}) : i_{e_1} = 0\} \subset C_1(e_1) \times C^1(e_1).$$

Unless  $r_{e_1} = 0$ ,  $A_R(e_1)$  is everywhere transverse to  $K_0(e_1)$ . The rest of the proof is quite similar to that of case 1.

**Case 3.** The element  $e_2$  is a linear resistor and  $e_1$  is a capacitor. In this case, instead of (10) the following holds;

$$(11) \quad K_0(\mathcal{E}) = \{(i_{R_1}, i_{R_2}, i_{e_2}, v_{R_1}, v_{R_2}, v_{e_2}) : i_{R(\mathcal{L}_j)} = A_{R_j \mathcal{L}_j}^t i_{R(\mathcal{L}_j)}, \\ v_{R(\mathcal{L}_j)} = -A_{R_j \mathcal{L}_j} v_{(\mathcal{L}_j)}, v_{e_2} = -\tilde{A}_{21} v_{R(\mathcal{T}_1)} - \tilde{A}_{22} v_{R(\mathcal{T}_2)}\}$$

where  $A_{R_j \mathcal{L}_j}$  is determined by the following form,  $j=1, 2$ ;

$$(12) \quad B_j = \begin{bmatrix} R(\mathcal{L}_j) & L_j & R(\mathcal{T}_j) & C_j \\ 1 & & A_{R_j \mathcal{L}_j} & A_{R_j C_j} \\ & 1 & A_{L_j \mathcal{L}_j} & A_{L_j C_j} \end{bmatrix} \begin{bmatrix} R(\mathcal{L}_j) \\ L_j \end{bmatrix}$$

and  $\tilde{A}_{2j}$  is the submatrix of  $A_{2j}$  consisting of the column corresponding to resistive tree elements,  $j=1, 2$ . In other words,  $K_0(\mathcal{E})$  is the graph of the map,

$$F: K_0(\mathcal{E}_1) \times K_0(\mathcal{E}_2) \times C_1(e_2) \rightarrow C^1(e_2)$$

defined by

$$F(i_{R_1}, v_{R_1}, i_{R_2}, v_{R_2}, i_{e_2}) = -\tilde{A}_{21} v_{R(\mathcal{T}_1)} - \tilde{A}_{22} v_{R(\mathcal{T}_2)}.$$

While we have;

$$(13) \quad A_R(\mathcal{E}) = A_R(\mathcal{E}_1) \times A_R(\mathcal{E}_2) \times A_R(e_2)$$

where

$$A_R(e_2) = \{(i_{e_2}, v_{e_2}) : v_{e_2} = i_{e_2} \cdot r_{e_2}\}.$$

We assert that  $A_R(\mathcal{E})$  is everywhere transverse to  $K_0(\mathcal{E})$  unless  $r_{e_2} = 0$  and  $\tilde{A}_{21} = \tilde{A}_{22} = 0$ . This can be verified by the straightforward calculation similar to the proof of Theorem 1, or we can convince ourselves by observing Fig. 1.



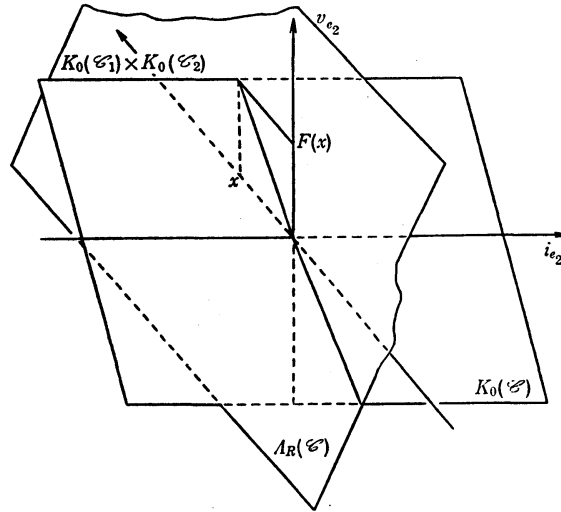


Fig. 1.

(The space  $K_0(\mathcal{E})$  is the rotated  $(i_{e_2}, K_0(\mathcal{E}_1) \times K_0(\mathcal{E}_2))$ -space with  $i_{e_2}$ -axis fixed. On the other hand, the space  $A_R(\mathcal{E})$  is the rotated  $(i_{e_2}, K_0(\mathcal{E}_1) \times K_0(\mathcal{E}_2))$ -space with  $K_0(\mathcal{E}_1) \times K_0(\mathcal{E}_2)$ -space fixed.) The rest of the proof is quite the same as before.

**Proof of Theorem 4.** According to Theorem A, we will show that for each  $(i_L, v_C) \in C_1(G_L(\mathcal{E})) \times C^1(G_C(\mathcal{E}))$  the space  $\pi_R(K(i_L, v_C))$  is transverse to  $A_R$ .

**Case 1.** The element  $e_1$  is a capacitor and  $e_2$  is an inductor. As before, the following holds;

$$(14) \quad A_R(\mathcal{E}) = A_R(\mathcal{E}_1) \times A_R(\mathcal{E}_2)$$

and

$$(15) \quad K_0(\mathcal{E}) = K_0(\mathcal{E}_1) \times K_0(\mathcal{E}_2).$$

For a point  $(i_R, v_R) \in \pi_R(K(i_L, v_C)) \cap A_R(\mathcal{E})$ , write

$$(i_R, v_R) = (i_{R_1}, i_{R_2}, v_{R_1}, v_{R_2})$$

and

$$i_{R_j} = (i_{R(\mathcal{I}_j)}, i_{R(\mathcal{L}_j)}), \quad v_{R_j} = (v_{R(\mathcal{I}_j)}, v_{R(\mathcal{C}_j)}), \quad j=1, 2.$$

Then the fact that  $(i_R, v_R)$  belongs to  $\pi_R(K(i_L, v_C))$  is equivalent to the following:

$$\begin{aligned} i_{R(\mathcal{I}_j)} &= A_{R_j \mathcal{I}_j}^t i_{R(\mathcal{L}_j)} + A_{L_j \mathcal{I}_j}^t i_{L_j} + \tilde{A}_{2j}^t i_{e_2}, \\ v_{R(\mathcal{C}_j)} &= -A_{R_j \mathcal{C}_j} v_{R(\mathcal{I}_j)} - A_{R_j \mathcal{C}_j} v_{C_j}. \end{aligned}$$

The assumption of  $L$ -connectedness implies the existence of inductor currents  $i'_{L_j} \in C_1(G_L(\mathcal{E}_j))$  such that

$$\mathbf{i}_{R(\mathcal{I}_j)} = A_{R_j \mathcal{I}_j}^t \mathbf{i}_{R(\mathcal{I}_j)} + A_{L_j \mathcal{I}_j}^t \mathbf{i}'_{L_j}, \quad j=1, 2.$$

This in turn means that the point  $(\mathbf{i}_{R_j}, \mathbf{v}_{R_j}) = (\mathbf{i}_{R(\mathcal{I}_j)}, \mathbf{i}_{R(\mathcal{I}_j)}, \mathbf{v}_{R(\mathcal{I}_j)}, \mathbf{v}_{R(\mathcal{I}_j)})$  belongs to  $\pi_R(K(\mathbf{i}'_{L_j}, \mathbf{v}_{C_j}))$ ,  $j=1, 2$ . Thus we have,

$$(\mathbf{i}_{R_j}, \mathbf{v}_{R_j}) \in \pi_{R_j}(K(\mathbf{i}'_{L_j}, \mathbf{v}_{C_j})) \cap \Lambda_R(\mathcal{E}_j), \quad j=1, 2.$$

By the assumption,  $T_{(\mathbf{i}_{R_j}, \mathbf{v}_{R_j})} \Lambda_R(\mathcal{E}_j)$  is transverse to  $\pi_{R_j}(K(\mathbf{i}'_{L_j}, \mathbf{v}_{C_j}))$  at  $(\mathbf{i}_{R_j}, \mathbf{v}_{R_j})$  in  $C_1(G_R(\mathcal{E}_j)) \times C^1(G_R(\mathcal{E}_j))$ ,  $j=1, 2$ . Since  $T_{(\mathbf{i}_{R_j}, \mathbf{v}_{R_j})}(\pi_{R_j}(K(\mathbf{i}'_{L_j}, \mathbf{v}_{C_j}))) = K_0(\mathcal{E}_j)$ ,  $j=1, 2$ , noting (14) and (15) we see  $T_{(\mathbf{i}_R, \mathbf{v}_R)} \Lambda_R$  is transverse to  $T_{(\mathbf{i}_R, \mathbf{v}_R)}(\pi_R(K(\mathbf{i}_L, \mathbf{v}_C))) (=K_0(\mathcal{E}))$  at  $(\mathbf{i}_R, \mathbf{v}_R)$ .

**Case 2.** The element  $e_1$  is a linear resistor and  $e_2$  is an inductor. In this case,  $\pi_R(\Lambda)$  and  $K_0(\mathcal{E})$  have the forms (9) and (10), respectively. For a point  $(\mathbf{i}_R, \mathbf{v}_R) \in \pi_R(K(\mathbf{i}_L, \mathbf{v}_C)) \cap \Lambda_R(\mathcal{E})$ , write

$$(\mathbf{i}_R, \mathbf{v}_R) = (\mathbf{i}_{R_1}, \mathbf{i}_{R_2}, \mathbf{i}_{e_1}, \mathbf{v}_{R_1}, \mathbf{v}_{R_2}, \mathbf{v}_{e_1})$$

and

$$\mathbf{i}_{R_j} = (\mathbf{i}_{R(\mathcal{I}_j)}, \mathbf{i}_{R(\mathcal{I}_j)}), \quad \mathbf{v}_{R_j} = (\mathbf{v}_{R(\mathcal{I}_j)}, \mathbf{v}_{R(\mathcal{I}_j)}), \quad j=1, 2.$$

since  $(\mathbf{i}_R, \mathbf{v}_R) \in \pi_R(K(\mathbf{i}_L, \mathbf{v}_C))$ , the following holds;

$$\begin{aligned} \mathbf{i}_{R(\mathcal{I}_j)} &= A_{R_j \mathcal{I}_j}^t \mathbf{i}_{R(\mathcal{I}_j)} + A_{L_j \mathcal{I}_j}^t \mathbf{i}_{L_j} + \tilde{A}_{2j}^t \mathbf{i}_{e_2}, \quad j=1, 2, \\ \mathbf{i}_{e_1} &= -\mathbf{i}_{e_2}, \\ \mathbf{v}_{R(\mathcal{I}_j)} &= -A_{R_j \mathcal{I}_j} \mathbf{v}_{R(\mathcal{I}_j)} - A_{R_j C_j} \mathbf{v}_{C_j}, \quad j=1, 2. \end{aligned}$$

Then the proof proceeds in the same way as in case 1.

**Case 3.** The element  $e_1$  is a capacitor and  $e_2$  is a linear resistor. In this case,  $\Lambda_R(\mathcal{E})$  and  $K_0(\mathcal{E})$  have the forms (13) and (11), respectively. For a point  $(\mathbf{i}_R, \mathbf{v}_R) \in \pi_R(K(\mathbf{i}_L, \mathbf{v}_C)) \cap \Lambda_R(\mathcal{E})$ , write

$$(\mathbf{i}_R, \mathbf{v}_R) = (\mathbf{i}_{R_1}, \mathbf{i}_{R_2}, \mathbf{i}_{e_2}, \mathbf{v}_{R_1}, \mathbf{v}_{R_2}, \mathbf{v}_{e_2})$$

and

$$\mathbf{i}_{R_j} = (\mathbf{i}_{R(\mathcal{I}_j)}, \mathbf{i}_{R(\mathcal{I}_j)}), \quad \mathbf{v}_{R_j} = (\mathbf{v}_{R(\mathcal{I}_j)}, \mathbf{v}_{R(\mathcal{I}_j)}), \quad j=1, 2.$$

The condition that  $(\mathbf{i}_R, \mathbf{v}_R) \in \pi_R(K(\mathbf{i}_L, \mathbf{v}_C))$  is equivalent to the following;

$$\begin{aligned} \mathbf{i}_{R(\mathcal{I}_j)} &= A_{R_j \mathcal{I}_j}^t \mathbf{i}_{R(\mathcal{I}_j)} + A_{L_j \mathcal{I}_j}^t \mathbf{i}_{L_j} + \tilde{A}_{2j}^t \mathbf{i}_{e_2}, \quad j=1, 2, \\ \mathbf{v}_{R(\mathcal{I}_j)} &= -A_{R_j \mathcal{I}_j} \mathbf{v}_{R(\mathcal{I}_j)} - A_{R_j C_j} \mathbf{v}_{C_j}, \\ \mathbf{v}_{e_2} &= -A_{21} \mathbf{v}_{\mathcal{I}_1} - A_{22} \mathbf{v}_{\mathcal{I}_2} + \mathbf{v}_{e_1}, \end{aligned}$$

where

$$\mathbf{v}_{\mathcal{I}_j} = (\mathbf{v}_{R(\mathcal{I}_j)}, \mathbf{v}_{C_j}), \quad j=1, 2.$$

The rest of proof proceeds as before.

4. Examples.

In our Theorems, we have treated the case of connection by different kinds of elements. If  $e_1$  and  $e_2$  are both capacitors or both inductors, then the new circuit may have no proper tree. If  $e_1$  and  $e_2$  are both resistors, Theorems 3 and 4 do not hold as shows the following example.

**Example 1.** Consider the circuits  $\mathcal{C}_1$  and  $\mathcal{C}_2$  shown in Fig. 2. Here the element  $R_1$  is non-linear current-controlled resistor having the characteristics shown in Fig. 4. The essential point is that the characteristic curve has a portion of negative inclination.

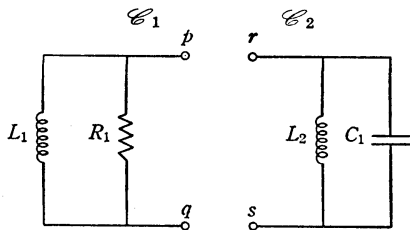


Fig. 2.

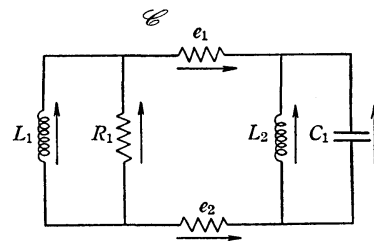


Fig. 3.

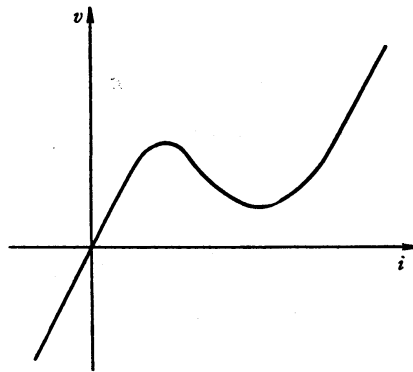


Fig. 4.

Connecting  $\mathcal{C}_1$  and  $\mathcal{C}_2$  by the resistors  $e_1$  and  $e_2$ , we obtain the new circuit  $\mathcal{C}$ . Although the circuit  $\mathcal{C}_1$  and  $\mathcal{C}_2$  are strongly well-posed, the new circuit  $\mathcal{C}$  is not necessary well-posed. Take the branches  $\mathcal{T}=\{R_1, e_1, C_1\}$  as a tree for  $\mathcal{C}$  and denote the fundamenta loop matrix associated with  $\mathcal{T}$  by  $B$ . Then we have,

$$B = \begin{bmatrix} L_1 & e_2 & L_2 & R_1 & e_1 & C_1 \\ 1 & 0 & 0 & -1 & 0 & 0 \\ 0 & 1 & 0 & -1 & -1 & 1 \\ 0 & 0 & 1 & 0 & 0 & -1 \end{bmatrix} \begin{matrix} L_1 \\ e_2 \\ L_2 \end{matrix}$$

$$B_R = \begin{bmatrix} e_2 & R_1 & e_1 \\ 1 & -1 & -1 \end{bmatrix} e_2,$$

$$Q_R = \begin{bmatrix} e_2 & R_1 & e_1 \\ 1 & 1 & 0 \\ 1 & 0 & 1 \end{bmatrix} \begin{matrix} R_1 \\ e_1 \end{matrix}.$$

Thus,  $(\dot{i}_R, \dot{v}_R) \in T_{(i_R, v_R)} K_0(\mathcal{C}) = K_0(\mathcal{C})$  if and only if

$$B_R \dot{v}_R = 0, \quad Q_R \dot{i}_R = 0.$$

While,  $(i_R, v_R) \in T_{(i_R, v_R)} A_R$  if and only if

$$R \begin{bmatrix} \dot{i}_R \\ \dot{v}_R \end{bmatrix} = 0,$$

where

$$R = \begin{bmatrix} e_2 & R_1 & e_1 & e_2 & R_1 & e_1 \\ -r_2 & 0 & 0 & 1 & 0 & 0 \\ 0 & -f'(i_{R_1}) & 0 & 0 & 1 & 0 \\ 0 & 0 & -r_1 & 0 & 0 & 1 \end{bmatrix} \begin{matrix} A_{e_2} \\ A_{R_1} \\ A_{e_1} \end{matrix}.$$

Therefore  $T_{(i_R, v_R)} A_R$  is transverse to  $K_0(\mathcal{C})$  if and only if the enlarged matrix

$$J = \begin{bmatrix} Q & 0 \\ 0 & B \\ & R \end{bmatrix}$$

has the full rank  $2\rho$ . By elementary operations, we can see  $\text{rank } J < 2\rho$  if and only if

$$\det \begin{bmatrix} 1 & 1 \\ r_2 + r_1 & -f'(i_{R_1}) \end{bmatrix} = 0$$

i.e.,

$$f'(i_{R_1}) = -(r_1 + r_2).$$

Thus, this is a counterexample for Theorems 3 and 4. Next, we will give some examples of strongly well-posed circuits.

### Example 2. (Linear circuits)

Any circuit whose resistors are all linear, positive and uncoupled is strongly well-posed. Each resistor  $R_j$  has the characteristics;  $v_{R_j} = i_{R_j} \cdot r_j$ ,  $r_j > 0$ ,  $j = 1, 2, 3, \dots, 2\rho$ . Let  $\mathcal{T}$  be a tree,  $\mathcal{L}$  a link,  $R(\mathcal{T})$  tree resistors and  $R(\mathcal{L})$  link resistors. Then the fundamental loop matrix  $B$  is decomposed as follows;

$$B = \begin{bmatrix} R(\mathcal{L}) & \mathcal{L}_{LC} & R(\mathcal{T}) & \mathcal{T}_{LC} \\ 1 & 0 & F_1 & F_2 \\ 0 & 1 & F_3 & F_4 \end{bmatrix} \begin{matrix} R(\mathcal{L}) \\ \mathcal{L}_{LC} \end{matrix}.$$

And

$$B_R = \begin{bmatrix} R(\mathcal{L}) & R(\mathcal{T}) \\ 1 & F_1 \end{bmatrix} R(\mathcal{L}).$$

Thus the enlarged matrix  $J$  is given by

$$J = \begin{bmatrix} R(\mathcal{L}) & R(\mathcal{T}) & R(\mathcal{L}) & R(\mathcal{T}) \\ -F_1^t & 1 & 0 & 0 \\ 0 & 0 & 1 & F_1 \\ 0 & 0 & 0 & -r_{\mathcal{T}}^{-1} \\ -r_{\mathcal{L}} & 0 & 1 & 0 \end{bmatrix} \begin{matrix} R(\mathcal{T}) \\ R(\mathcal{L}) \\ R(\mathcal{T}) \\ R(\mathcal{L}) \end{matrix},$$

where  $r_{\mathcal{T}}$  and  $r_{\mathcal{L}}$  are diagonal matrices with positive entries. The matrix  $J$  has the full rank if and only if

$$\det \begin{bmatrix} -F_1^t & r_{\mathcal{T}}^{-1} \\ r_{\mathcal{L}} & F_1 \end{bmatrix} \neq 0.$$

This is equivalent to  $\det [F_1 F_1^t + r_{\mathcal{L}} r_{\mathcal{T}}^{-1}] \neq 0$ . And this is always assumed because  $F_1^t F_1$  is positive semi-definite and  $r_{\mathcal{L}} r_{\mathcal{T}}^{-1}$  is positive definite.

Of course, if the resistors are not linear but monotone increasing, then the result remains valid. More generally, the following holds.

**Example 3.** If the characteristics of the resistors have the following form with respect to some tree,

$$A_R = A_R(\mathcal{T}) \times A_R(\mathcal{L})$$

and

$$A_R(\mathcal{T}) = \{(i_{\mathcal{T}}, v_{\mathcal{T}}) : i_{\mathcal{T}} = f(v_{\mathcal{T}})\}, \quad A_R(\mathcal{L}) = \{(i_{\mathcal{L}}, v_{\mathcal{L}}) : v_{\mathcal{L}} = g(i_{\mathcal{L}})\}$$

and if at each point  $(i_R, v_R) \in A_R$

$$\det \begin{bmatrix} -F_1^t & f'(v_{\mathcal{T}}) \\ g'(i_{\mathcal{L}}) & F_1 \end{bmatrix} \neq 0.$$

then the circuit is strongly well-posed.

Next, we consider a rather old-fashioned electron tube amplifier.

**Example 4.** (A unit amplifier using a triode.)

Consider the circuit  $A_1$  shown in Fig. 5. The coupled resistors  $R_b$  and  $R_c$  consists a triode. The resistors  $R_g$  and  $R_k$  are uncoupled positive linear resistors. The resistor  $R_p$  is a linear resistor with a voltage source, its characteristics is represented as shown in Fig. 6. The electron tube  $\{R_b, R_c\}$  has the following characteristics;

$$A_{R_{b,c}} = \{(i_b, i_c, v_b, v_c) : i_b = f_b(v_b, v_c), i_c = f_c(v_b, v_c)\}$$

where the function  $f_b$  and  $f_c$  are supposed to be smooth and at each point to satisfy the condition;

$$\frac{\partial f_b}{\partial v_b} > 0, \quad \frac{\partial f_c}{\partial v_c} > 0.$$

This is satisfied at least within the region usually used. A typical form of  $f_b$  and  $f_c$  are illustrated as follows.

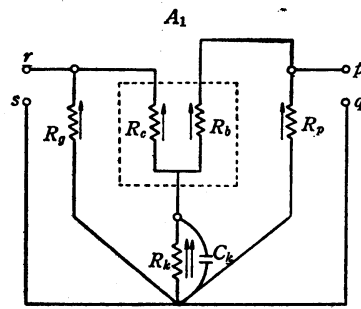


Fig. 5.

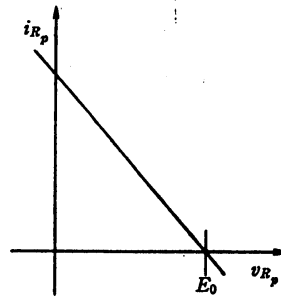


Fig. 6.

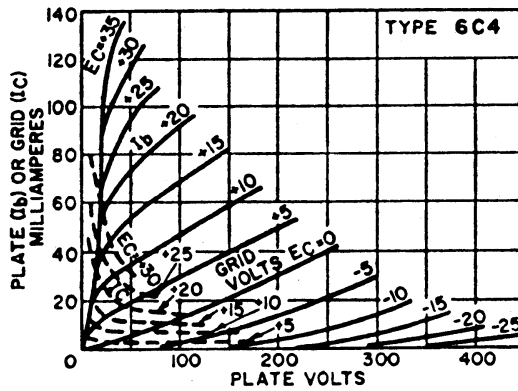


Fig. 7. (From RCA tube manual.)

If we take  $\{C_k, R_b, R_c\}$  as a tree, then the condition of example 3 is satisfied. For the condition

$$\det \begin{bmatrix} 0 & 0 & 1 & \frac{\partial f_b}{\partial v_b} & \frac{\partial f_b}{\partial v_c} \\ 1 & 0 & 1 & \frac{\partial f_c}{\partial v_b} & \frac{\partial f_c}{\partial v_c} \\ r_g & 0 & 0 & 0 & -1 \\ 0 & r_k & 0 & 0 & 0 \\ 0 & 0 & r_p & -1 & 0 \end{bmatrix} = 0$$

is equivalent to

$$1 + \frac{\partial f_b}{\partial v_b} \cdot r_p + \frac{\partial f_c}{\partial v_c} \cdot r_g = 0.$$

Hence for any  $r_p > 0$  and  $r_g > 0$ , the circuit is strongly well-posed.

Applying our results inductively, we obtain,

**Example 5.** (Triode amplifier)

Let  $A_i$  be a copy of the circuit  $A_1$ ,  $i=1, 2, 3, \dots, n$ . Connect  $A_i$ 's as shown in the following. Then this circuit is strongly well-posed.

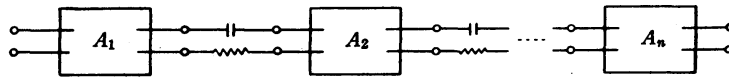


Fig. 8.

Now, consider a circuit for which we can take a tree consisting only capacitors, and we have,

$$B_r = \begin{bmatrix} 1 & & & & \\ & 1 & & & \\ & & 1 & & \\ & & & \ddots & \\ & & & & 1 \end{bmatrix}, \quad J = \begin{bmatrix} 0 & 1 & & & \\ & & 1 & & \\ & & & \ddots & \\ & & & & 1 \\ D_{i_R} g & & & & D_{v_R} g \end{bmatrix}$$

where

$$A_R = \{(\mathbf{i}_R, \mathbf{v}_R) : g(\mathbf{i}_R, \mathbf{v}_R) = 0\}.$$

Therefore the circuit is strongly well-posed provided  $\det D_{i_R} g \neq 0$ . This argument shows the usual transistor flip-flop circuit with stray capacitors as shown in Fig. 9 is strongly well-posed.

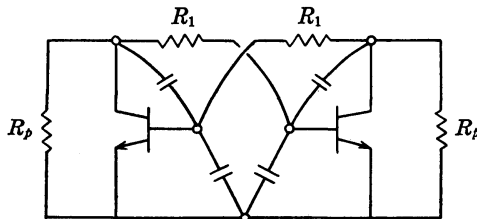


Fig. 9.

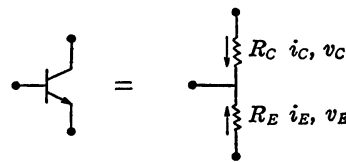


Fig. 10.

The characteristics of the transistor is given by

$$\begin{aligned} i_C &= f_C(v_C) - \alpha_{21} i_E, \\ i_E &= f_E(v_E) - \alpha_{12} i_C, \end{aligned}$$

where

$$0 < \alpha_{12}, \alpha_{21} < 1.$$

Since  $R_1$  and  $R_p$  are linear uncoupled (rigorously saying,  $R_p$  is affine, for  $R$  contains a voltage source), it is sufficient to verify the determinant of the following  $2 \times 2$  matrix is non-zero,

$$\begin{array}{cc} i_C & i_E \\ \left[ \begin{array}{cc} 1 & \alpha_{21} \\ \alpha_{12} & 1 \end{array} \right] \begin{array}{l} R_C \\ R_E \end{array} \end{array}.$$

Since  $0 < \alpha_{12}, \alpha_{21} < 1$ , the circuit is strongly well-posed.

More generally, we have.

**Example 6.** Any transistor circuit consisting of transistors, uncoupled strictly monotone resistors, capacitors and inductors having a tree consisting of only capacitors is strongly well-posed.

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